Continuous wavelet transform

In this section we describe step-by-step our empirical wavelet mode decomposition. First, since the raw signals consist in unevenly spaced time series, each dataset is re-sampled (using linear interpolations when necessary) at the frequency of one observation every ten years and centered so that it has zero mean.

The Lomb-Scargle periodograms plotted in Fig.5 (see text) are obtained with the plomb Matlab function from the raw (irregularly spaced) signals.

The rest of this work is based on the continuous wavelet transform (CWT); self-made codes implemented for Scilab are available upon request (see also the toolbox ...). The main wavelet-related principles involved in this study are inspired from (Deliège et al., 2016; Nicolay S., 2011; Nicolay et al., 2009)] and are summarized below. More information about wavelets and the CWT can be found in e.g. (Daubechies I., 1992; Mallat S., 1999; Meyer Y., 1993; Torrence and Compo, 199).

Given a wavelet ψ and a function f, the wavelet transform of f at time t and at scale a > 0 is defined as

$$W_f(t,a) = \int f(x)\bar{\psi}\left(\frac{x-t}{a}\right)\frac{dx}{a}$$

where ψ is the complex conjugate of ψ . In the context of time-frequency analysis, it is recommended to use a wavelet well-located in the frequency domain ([7]). In this work, we use the wavelet ψ defined by its Fourier transform as

$$\hat{\psi}(\nu) = \sin\left(\frac{\pi\nu}{2\Omega}\right) e^{\frac{-(\nu-\Omega)^2}{2}}$$

with $\Omega = \pi \sqrt{2}/\ln 2$, which is similar to the Morlet wavelet but with exactly one vanishing moment ([5]). Since $|\hat{\psi}(\nu)| < 10^{-5}$ if $\nu \le 0$, ψ can be considered as a progressive wavelet (i.e. is zero for negative arguments) and thus allows an easy recovery of trigonometric functions. Indeed, if $f(x) = \cos(\omega x)$, then

$$W_f(t,a) = \frac{1}{2} \mathrm{e}^{it\omega} \overline{\hat{\psi}(aw)}$$

Therefore, for a fixed time t, if a^* stands for the scale at which the function

 $a \rightarrow |W_f(t,a)|$ reaches its maximum, then we have $a^*\omega = \Omega$. Consequently, the value of ω can be obtained (if unknown) and *f* is recovered with the real part of its CWT: $f(x) = 2\Re(W_f(x,a^*))$.

Drawing on this principle, the general scheme for extracting the dominant components of a given function f is the following.

1. The CWT of *f* is computed and the application $(t,a) \rightarrow |W_f(t,a)|$ can be seen as a time-frequency representation of *f*.

2. The wavelet-based spectral composition of f, also called wavelet spectrum of f, i.e. the wavelet counterpart of its Fourier spectrum, is obtained as

$$\Lambda: a \to E_t |W_f(t,a)|,$$

where E_t denotes the mean over time.

3. For each scale *s* at which Λ reaches a local maximum, a component c_s is extracted from *f* as $c_s(x) = 2\Re(W_f(x,s))$.

By adding the modes obtained this way, the CWT allows an almost complete reconstruction of f with smooth amplitude modulated-frequency modulated (AM-FM) components. However, it is slightly different in practice because the signal considered has a finite length and thus has to be padded at its edges. The chosen padding (zero-padding in the present case) irremediably flaws a certain proportion of the wavelet coefficients, which creates a so-called cone of influence (see e.g. Deliège and Nicolay, 2016). Even though it is generally impossible to correct exactly the border effects, noticeable improvements can be made to sharpen the edges of the components extracted and the reconstruction of the initial signal by iterating the process described above. More precisely, the technique used in this work in complement to the previous algorithm is the following.

Let us imagine that the components of interest are located at scales $(s_i)_{i \in I}$ (for some set of indices *I*). Then, at the first iteration of steps 1 to 3 we extract $(c_i^{-1})_{i \in I}$ as $c_i^{-1}(x) = 2\Re(W_f(x,s_i))$. We now perform the CWT and the extraction at the same scales $(s_i)_{i \in I}$ but with

$$f_1 = f - \sum_{i \in I} c_i^{\perp}$$

and get the modes $(c^2_i)_{i \in I}$, then again repeat the process with $f_2 = f_1 - \sum_{i \in I} c^2_i$, and so on. The iterations stop when the components extracted are not significant anymore, i.e. at iteration *J* if $\max \{ \|c_i^J\|, i \in I \} < \alpha \max \{ \|c_i^1\|, i \in I \}$

where $\|.\|$ denotes the energy (square of L^2 norm) of a signal and α is a threshold chosen as 0.01.

The final components $(c_i)_i \in I$ are then obtained as

$$C_i = \sum_{i=1}^{J} C^j_{i.}$$

Although most of the energy of *f* has already been drained at the end of the procedure described in this section, it may turn out useful to examine what kind of information still remains in *f*. For that purpose, the whole method can be applied to $f - \sum_{i \in I} c_i$ to find out which frequencies might have been hidden behind the dominant ones and thus obtain "secondary components", if any.

References

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Original data (PN trace element contents, sunspot number, and temperature) were filtered according to the dominant periods revealed by the CWT (Fig.1-5).



Figure 1: the Gleissberg cycle (70-100 yr)



Figure 2: de Vries cycle (200-210 yr)



Figure 3: Unnamed cycles 350 yrs



Figure 4: Unnamed cycles 550 yrs



Figure 5: Eddy cycle (1000 yr)